

Introduction to
FMT Modulation and
Multiuser Multitone Wireless Uplink Systems

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Outline

❑ Filtered Multitone (FMT) Modulation

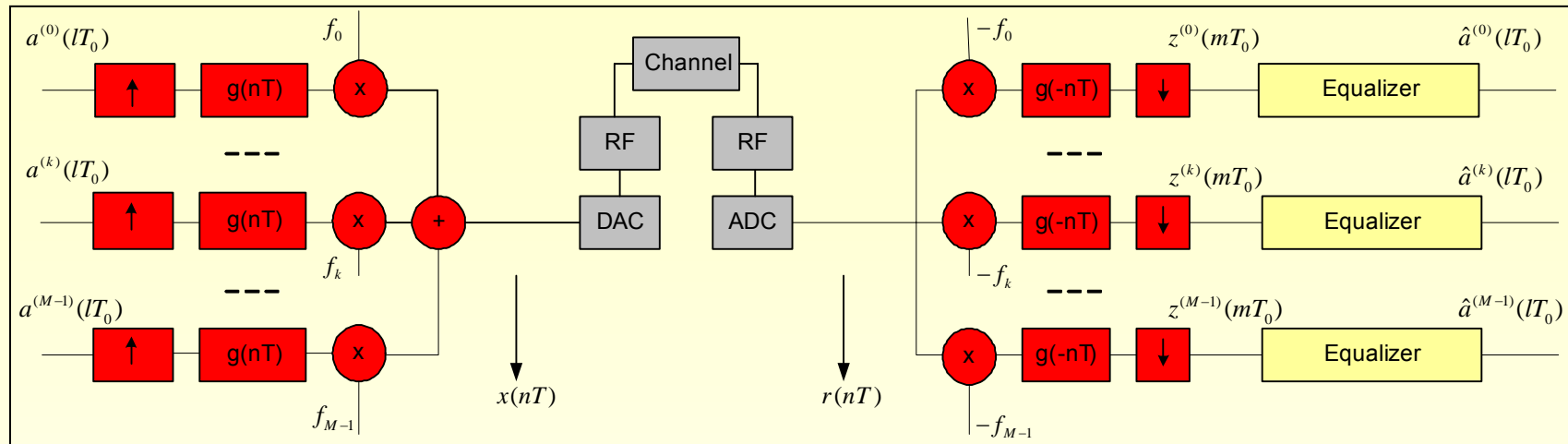
- Principles
- Detection and Performance Limits
- Synchronization

❑ Asynchronous Multiuser Multitone Systems

- Multiuser DMT/OFDM
- Multiuser FMT
- Synchronization in Multiuser FMT

FMT Modulation

Multicarrier Architecture

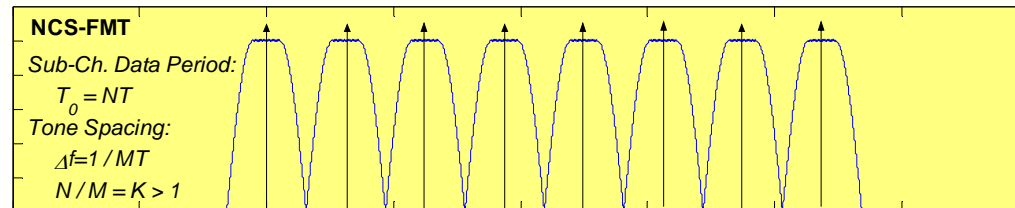
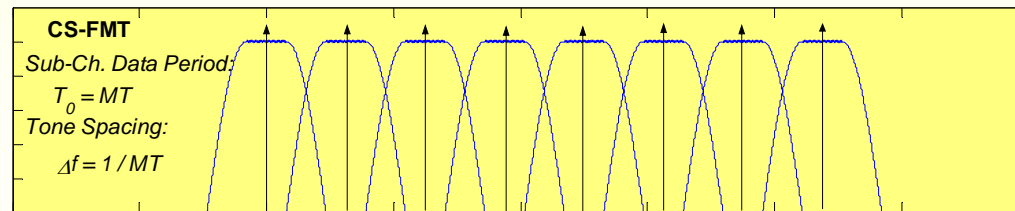
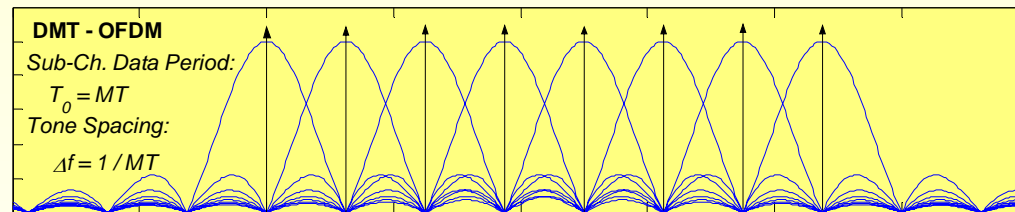


□ Transmit a high data rate signal through a number of low rate sub-channels.

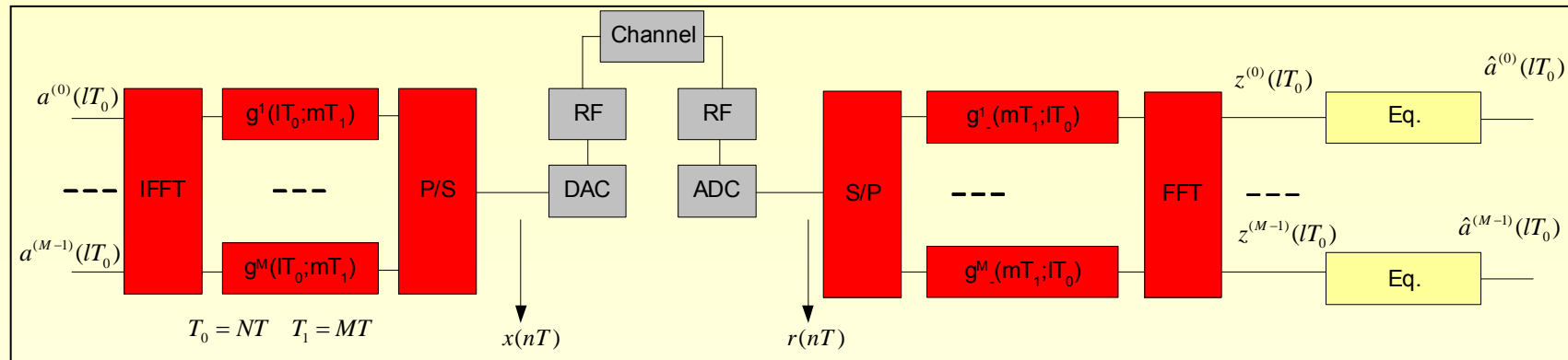
- ✓ **DMT** (Discrete Multitone): well known OFDM scheme. It deploys a prototype filter with rectangular impulse response.
- ✓ **FMT** (Filtered Multitone): deploys a sub-channel prototype pulse with time-frequency concentrated response.

Filtered Multitone Architecture

- ❑ TX bandwidth: $W = 1 / T$.
- ❑ Sub-carriers: $f_k = k/(MT)$, $k=0, \dots, M-1$.
- ❑ Sub-channel period: $T_0 = NT$
- ❑ **DMT – OFDM** : Rectangular impulse response prototype pulse $g(nT)$.
- ❑ **FMT** : Frequency concentrated prototype pulse, e.g., root-raised-cosine.

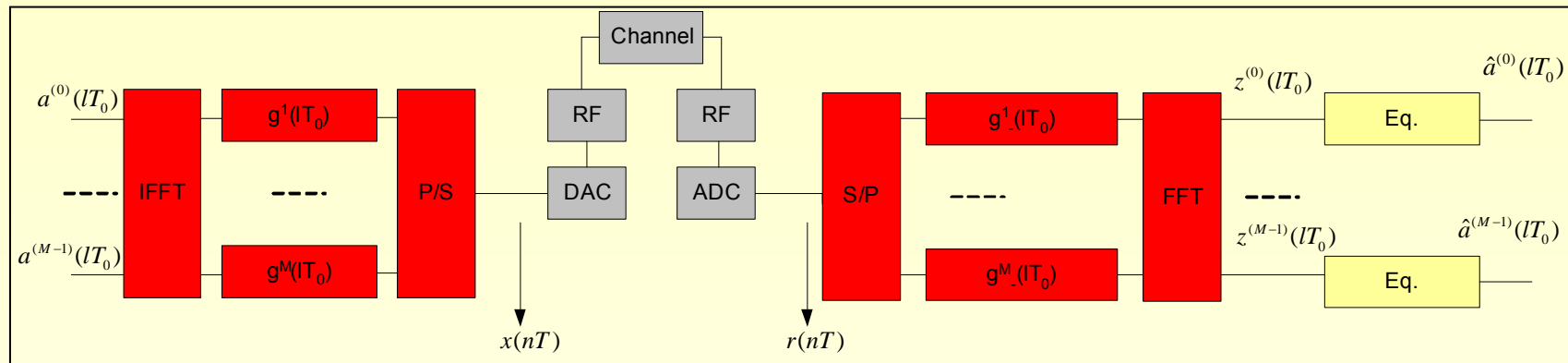


Non-Critically Sampled FMT Efficient Implementation



- ❑ A possible efficient implementation is based on FFT and low-rate filtering as proposed by *Cherubini, Eleftheriou, Olcer, Cioffi, Comm. Mag. 2000*.
- ❑ In NCS-FMT the sub-channel pulse is cyclically time-variant.

Critically Sampled FMT Efficient Implementation



CS-FMT

$$f_k = \frac{k-1}{T_0} \quad T_0 = MT \quad k = 1, \dots, M$$

$$g^k(mT_0) = g((k-1)T + mT_0) \quad m \in \mathbb{Z}$$

↓
Prototype pulse

Sub-channel Matched Filter Output

- ❑ RX front-end output sampled at symbol rate:

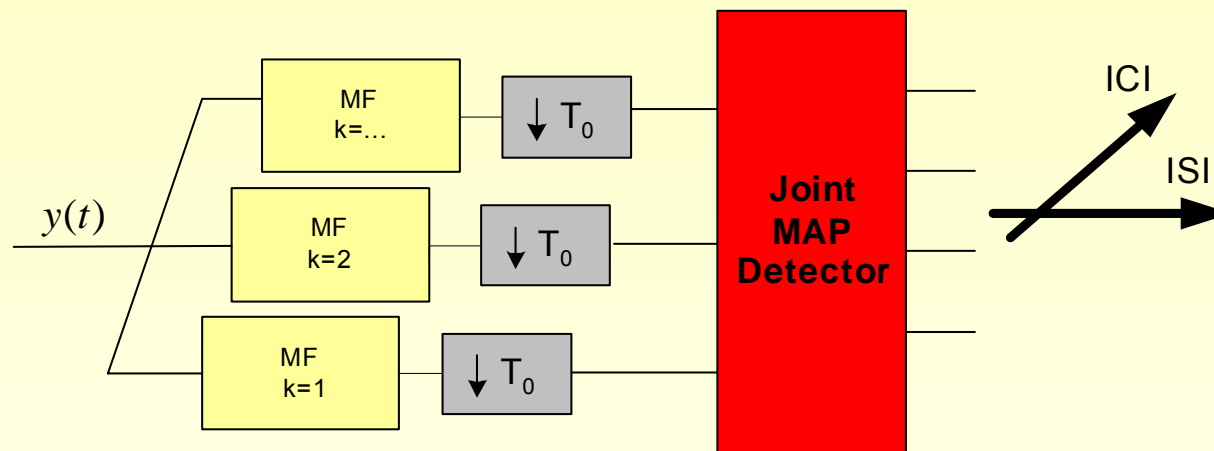
$$z^{(k)}(lT_0) = a^{(k)}(lT_0)g_{EQ}^{(k)}(0) + \sum_{m \neq 0} a^{(k)}(lT_0 - mT_0)g_{EQ}^{(k)}(mT_0) + ICI^{(k)} + \eta^{(k)}(lT_0)$$

- $\Delta t=0, \Delta f=0$: No ICI with band limited pulses but some ISI because of the channel frequency selectivity or non-ideal Nyquist pulses.
 - $\Delta t \neq 0, \Delta f=0$: Increased ISI because of wrong sampling phase.
 - $\Delta t=0, \Delta f \neq 0$: ISI and some ICI when Δf exceeds the frequency guards.
- ❑ When frequency concentrated pulses are deployed we get small ICI, however we need to run sub-channel equalization.
 - ❑ We need time/frequency synchronization to minimize the amount of ISI and ICI.

Detection in FMT

- ❑ FMT requires some form of equalization.
- ❑ The optimal detector is a multi-channel MLSE/MAP equalizer.
- ❑ Independent sub-channel equalization via linear or DFE equalization
 - ✓ As we increase the number of tones we obtain narrower sub-channels.
 - ✓ The sub-channel equalizer has low complexity since the sub-channel impulse response is short (sub-channel is narrow band).

Optimal FMT Detector

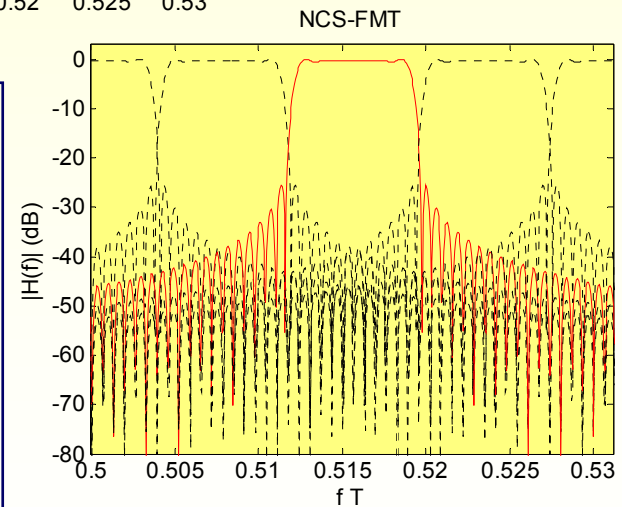
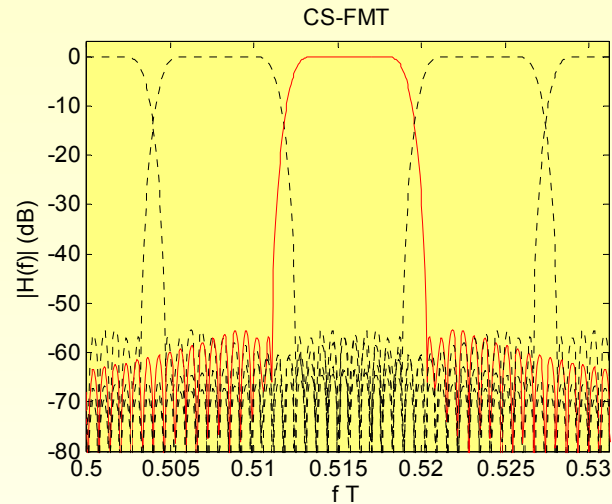
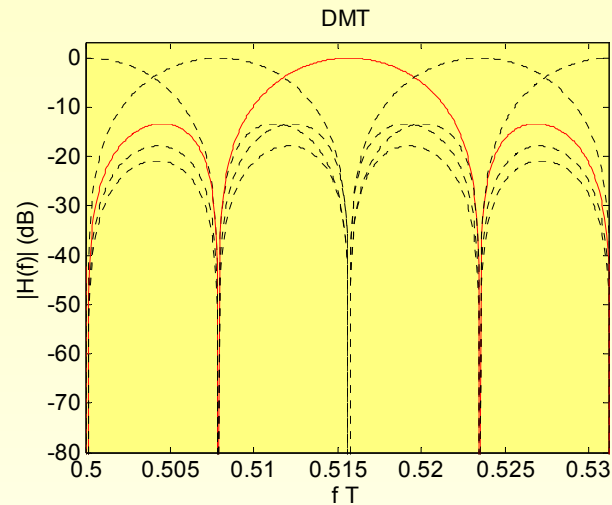


- The MAP detector computes the a posteriori probabilities of all data symbols by observing the input signal $y(t)$ over a time window.

Ref: Tonello, *BLTJ 2003, Proc. VTC 2002 Fall*.

Spectral Efficiency

Example of Sub-Channel Frequency Response



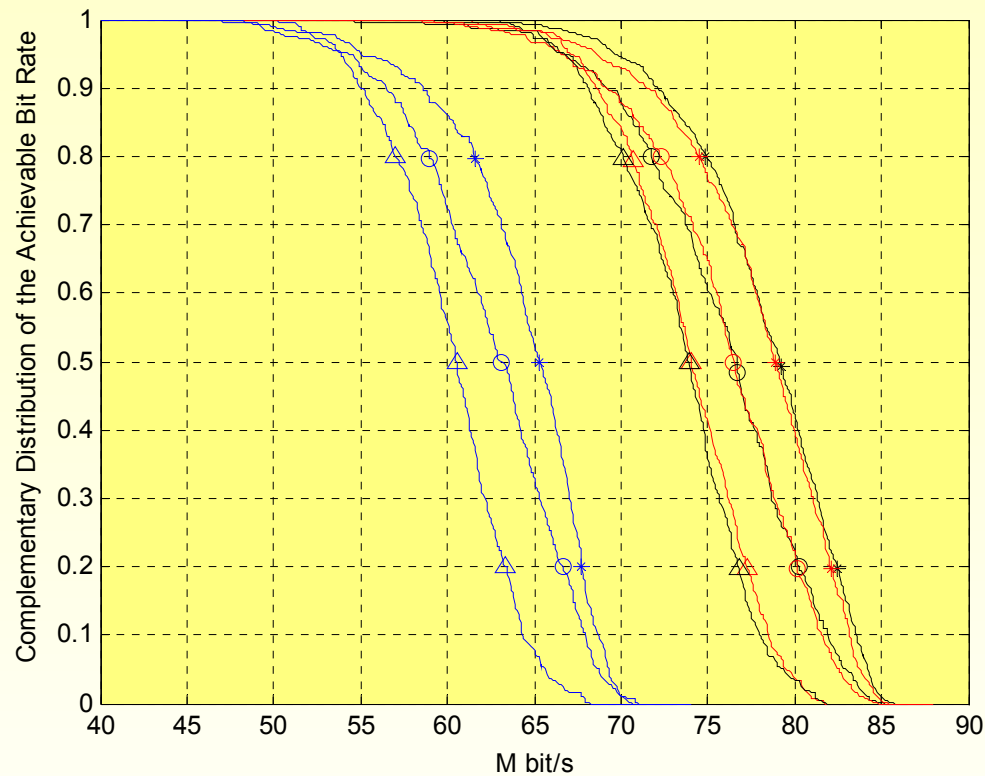
$M = 128$ $B = 25$ MHz

CS-FMT: rectangular windowed pulses + 4 virtual carriers

NCS-FMT: root raised cosine pulses $N/M = 1.125$ + 4 virtual carriers

CP-DMT: CP length = 30 chips + 16 virtual carriers

Probability [Achievable Bit Rate > K]



--- DMT
--- CS-FMT
--- NCS-FMT

▲ Rayleigh exponential with $\tau_{rms}=100$ ns
● Rayleigh exponential with $\tau_{rms}=40$ ns
* Ricean exponential with $R=5$ dB, $\tau_{rms}=40$ ns

- DFE sub-channel equalization in FMT.
- FMT has higher spectral efficiency than DMT/OFDM.

Performance Limits for FMT Modulation

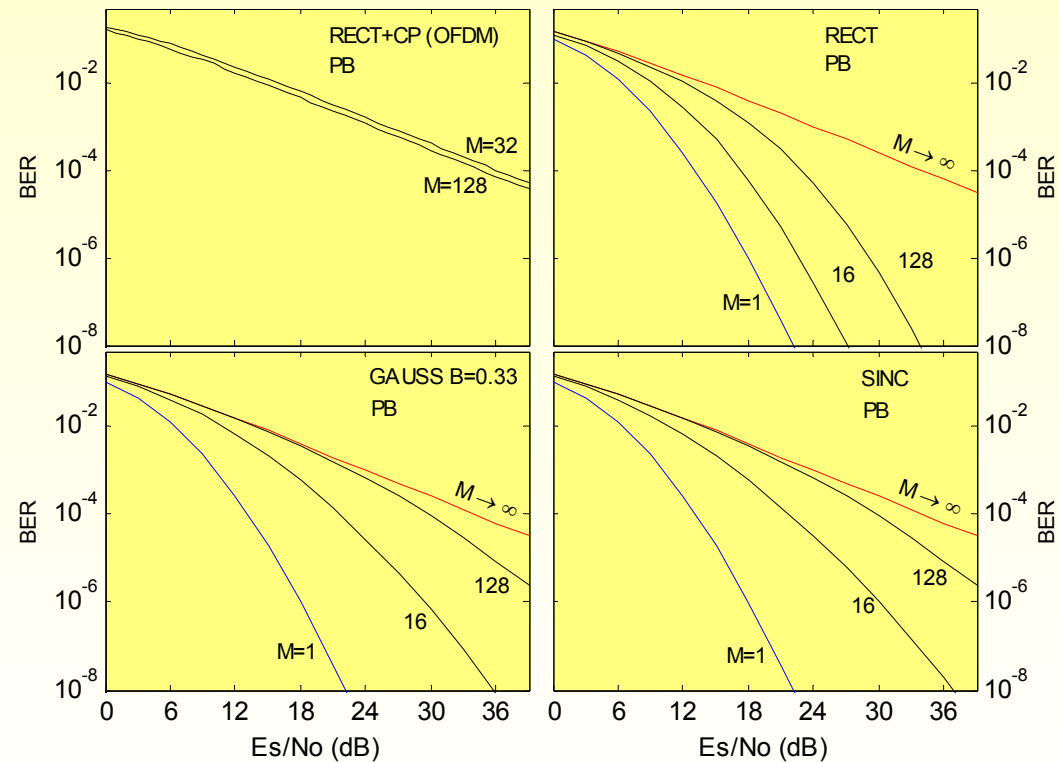
Performance Limits

- ❑ Let us consider FMT modulation in **time variant frequency selective fading**.
- ❑ We can evaluate the **matched filter bound performance** in closed form:
 - BER performance with an ideal equalizer
- ❑ The analysis yields very interesting insights:
 - FMT modulation is a diversity transform and offers
 - time-frequency diversity gains and coding gains that are a function of the prototype pulse, and number of tones.

Ref: Tonello, *IEEE Trans. on Wireless Comm. in press* ; Tonello, *Proc. WPMC 2003*.

Frequency Selective Time-Invariant Channel

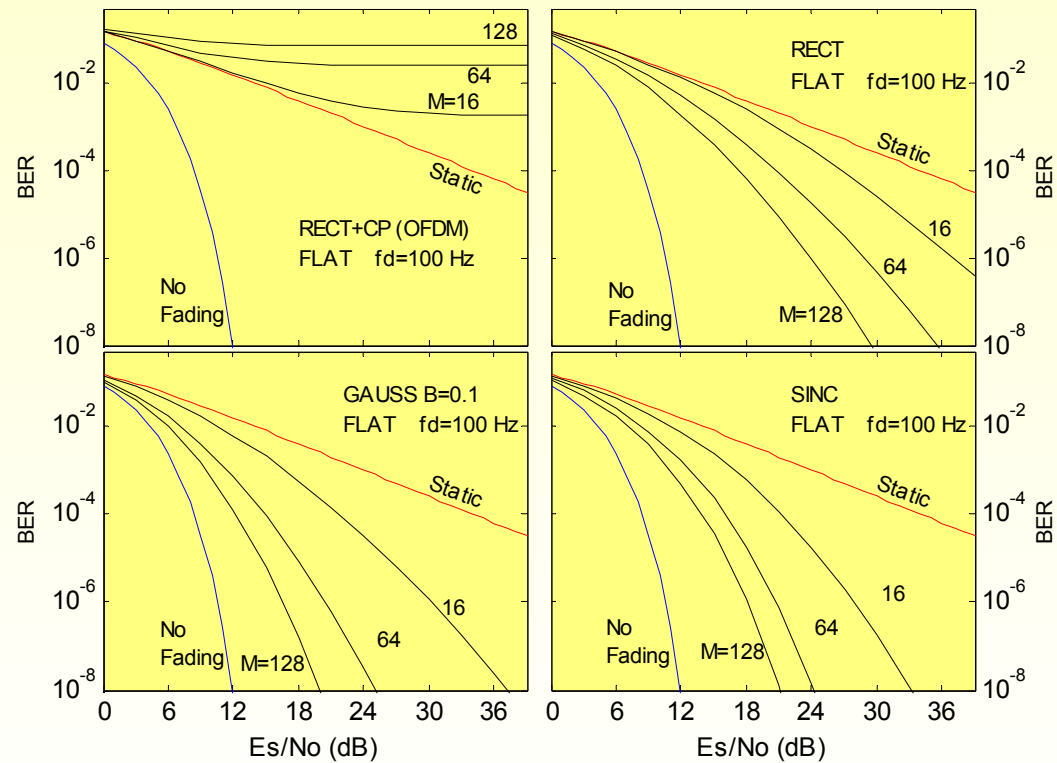
ITU PB channel with quasi static fading
Bandwidth $W=3.84$ MHz. BPSK modulation.



MF Bound performance for FMT with *rectangular*, *Gaussian*, and *sinc* prototype pulse

Time-Variant Channel

Flat fading with Jakes' Doppler spectrum
Bandwidth $W=24.3$ kHz. BPSK modulation.



MF Bound performance for FMT with *rectangular*, *Gaussian*, and *sinc* prototype pulse

Remarks

❑ In Frequency Selective Fading

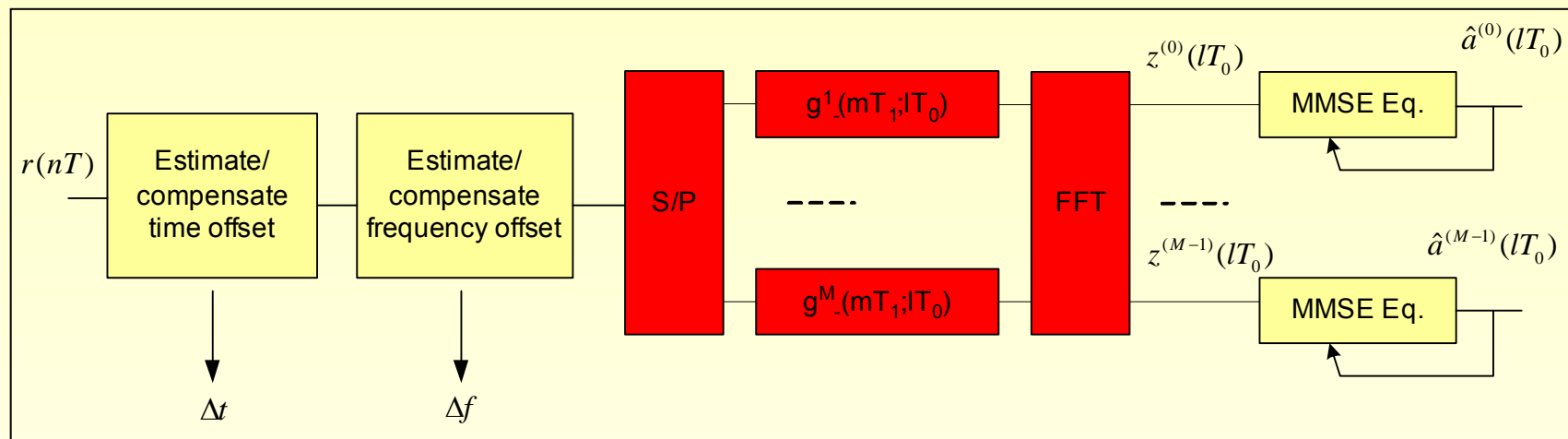
- ✓ FMT is a good choice complexity wise
- ✓ Single Carrier modulation is a good choice performance wise.

❑ In Time Selective Fading

- ✓ FMT is a good choice performance wise
- ✓ Single carrier modulation is a good choice complexity wise.

Synchronization

Time-Domain Synchronization



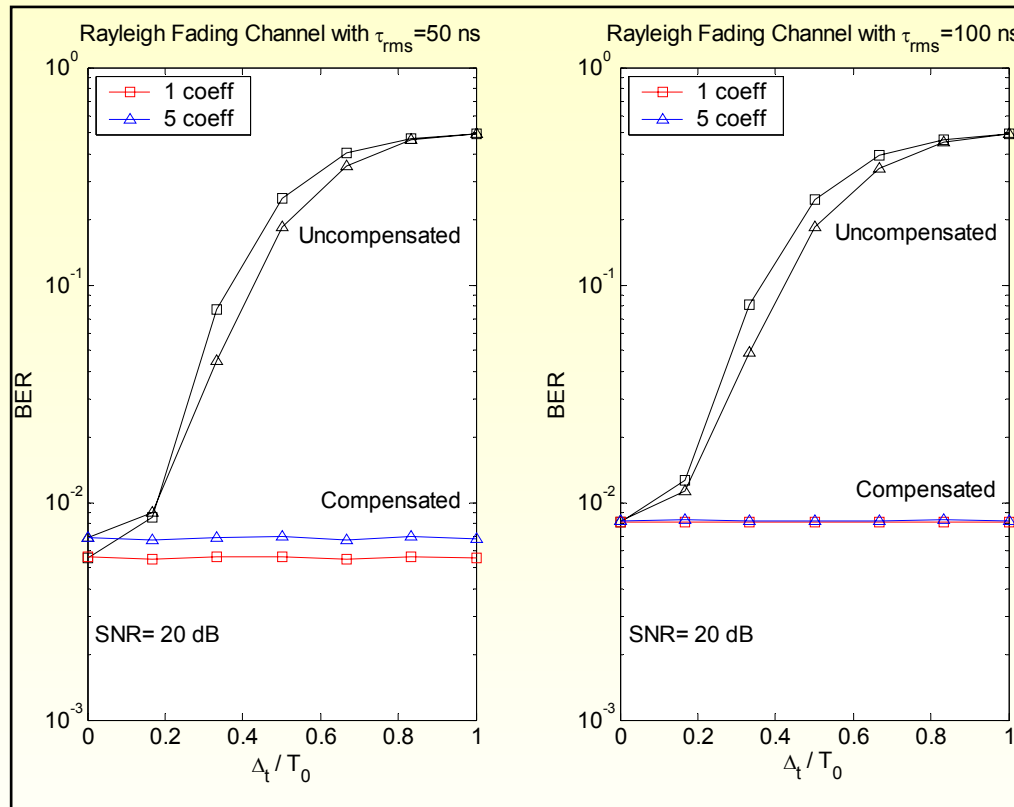
□ Estimation and compensation of time/frequency offsets can be done in the time domain.

- **Blind Synchronization.** The drawback is that channel estimation cannot rely on known training symbols.
- **Cyclic Training Approach.** We generate training sequences that exhibit a periodic behavior at the tx-rx side, similarly in spirit to *Schmidl and Cox* method in OFDM.
- **PN Training Approach.** We generate PN training sequences.

Ref: Assalini, Tonello, Proc. *WPMC 2003* ; Tonello, Rossi, Proc. *WPMC 2004*.

Bit-Error-Rate Performance with Time-Domain Sync.

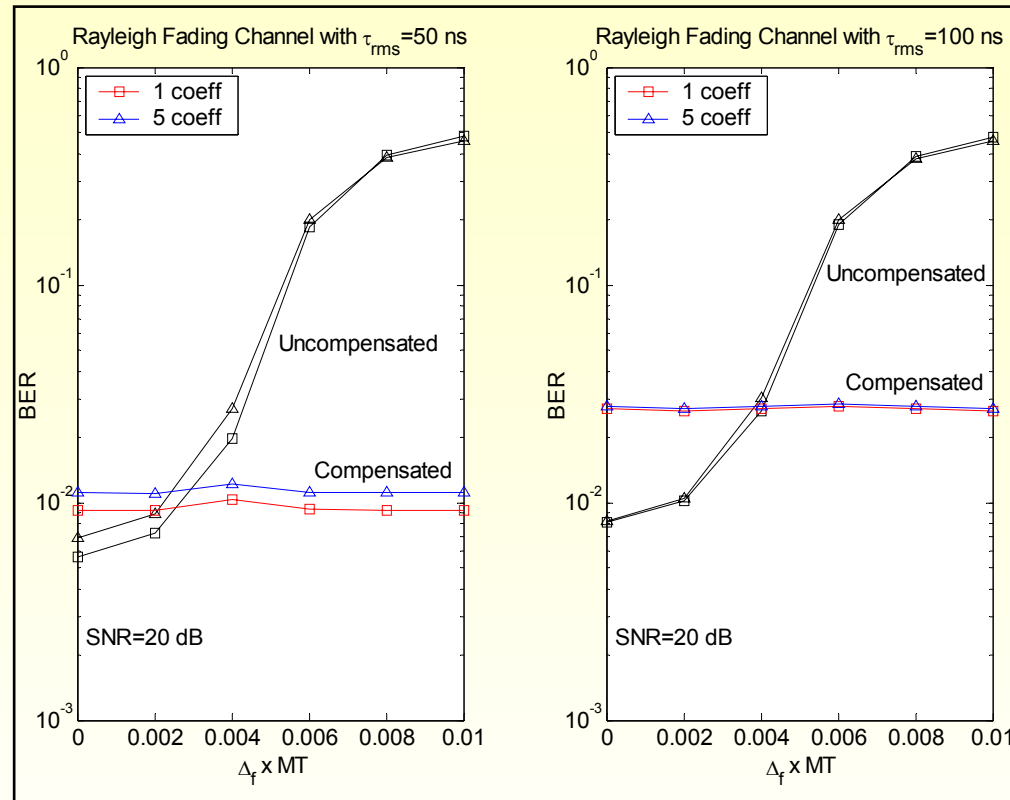
Timing Error



- 4-PSK modulation, M=32 Sub-channels, Root-Cosine Pulses, Bandwidth 20 MHz.
- Performance drastically decreases without time/frequency compensation.
- Synchronization with random sequences and 1-5 Tap RLS equalizer.

Bit-Error-Rate Performance Time-Domain Sync.

Frequency Error



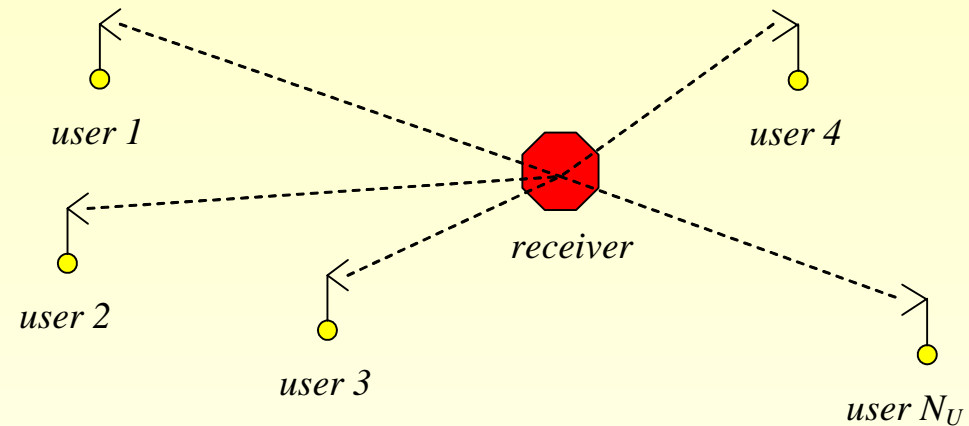
- The practical scheme with random training sequences performs well.
- A single tap equalizer is a good choice for delay spreads of ~ 100 ns.
- The performance of the synchronizer can be improved with a fine sync. method in the frequency domain.

Remarks

- ❑ OFDM is an elegant simple solution to cope with the channel frequency selectivity.
- ❑ FMT can yield higher spectral efficiency than DMT/OFDM.
- ❑ The sub-channel spectral containment of FMT makes it more robust to time and carrier frequency offsets than OFDM.
- ❑ Time-Frequency acquisition is still of great importance.
- ❑ FMT can be more complex than DMT since it requires filtering and equalization:
 - *we save complexity by using a smaller number of tones than in OFDM.*

Multiuser Multitone Architectures

Asynchronous Multiple Access Channel

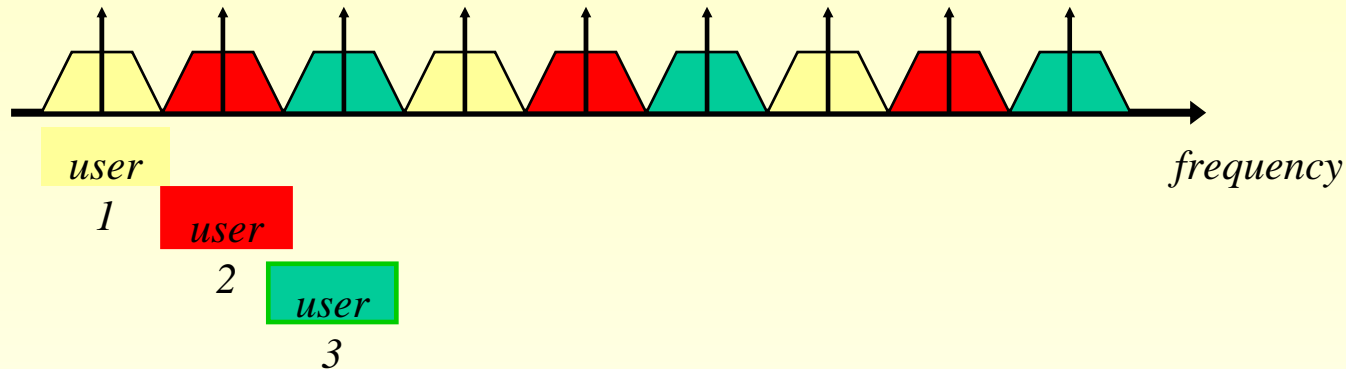


□ Consider the Uplink

- Users are asynchronous
- Time offsets between users (*propagation delays*)
- Carrier frequency offsets between users (*oscillators, Doppler*).

□ Frequency Division Multiplexing is an Interesting Solution.

Multitone Multiple Access



- M total tones
- Groups of tones are assigned to users
- Each user deploys multicarrier modulation over its set of tones
- Several tone allocation methods are possible:
 - ✓ *Orthogonal or non-orthogonal, Static or dynamic...*

Solutions

- Two efficient implementation solutions
 - ✓ Multiuser DMT : **MU-DMT** (OFDMA)
 - ✓ Multiuser FMT : **MU-FMT**

Ref: Tonello, Pupolin, Proc. *WPMC 2001* ; Tonello, *Bell Labs Tech. Journ.* 2003.

Received Composite Signal

$$y(t) = \sum_{u=1}^{N_U} \sum_{k=0}^{M-1} \sum_{l=-\infty}^{\infty} a^{(u,k)} (lT_0) g_R^{(u,k)} (t - \Delta t_{u,k} - lT_0; t) e^{j(2\pi\Delta f_{u,k}t + \phi_{u,k})} + \eta(t)$$

Time-variant channel response for user u and sub-channel k

Time offset of user u and sub-channel k due to the propagation delay

Carrier Frequency offset due to Doppler and oscillators precision

Propagation delay: $\Delta t = 6.6 \text{ us/km} = 6.6 \text{ chips/km}$ [B=1 MHz]

Doppler: $\Delta f = 0.93 \text{ Hz}$ [v=1 km/h -- fc=1 GHz]

Oscillator: $\Delta f = 1000 \text{ Hz}$ [fc=1 GHz -- 1 p.p.m]



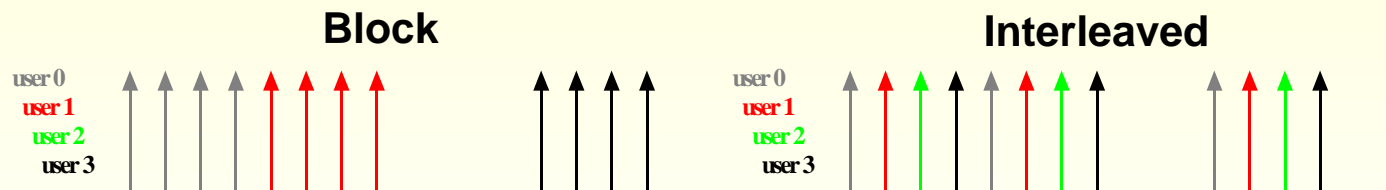
The values depend upon the hardware and the mobility/coverage requirements

MAI-ICI-ISI Components

The interference depends upon

- sub-carrier spacing
- prototype filter
- tone allocation

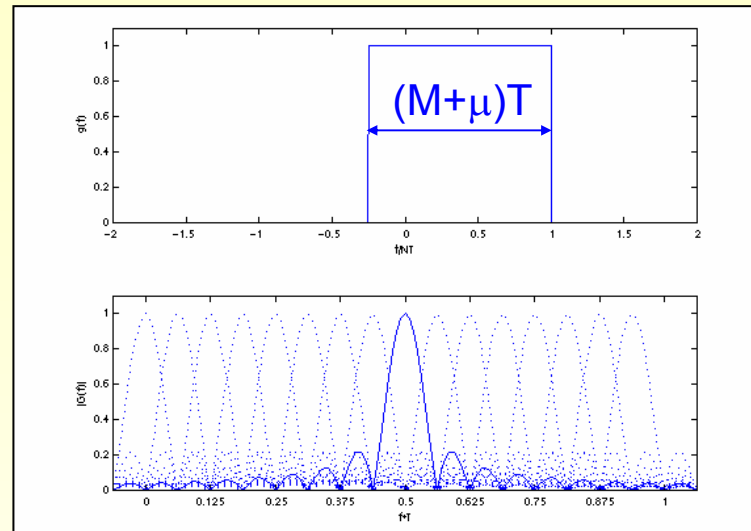
Tone Allocation



- **Block allocation** yields lower MAI than the interleaved.
- **Tone interleaving** is a better option to exploit the frequency diversity.
- **Frequency guards** can be inserted only with the block allocation in OFDM.
- The NCS-FMT architecture can comprise frequency guards.

MU-OFDM Uplink Analysis

Interference in MU-OFDM

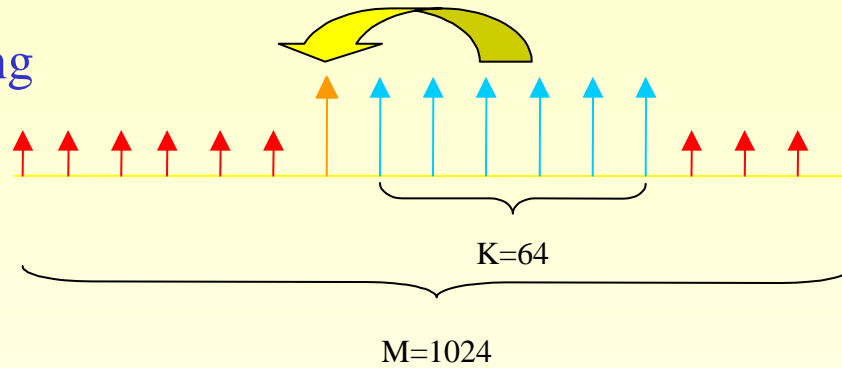


- ❑ The system is orthogonal in an ideal synchronous channel.
 - Users' time/frequency offsets generate MAI.
- ❑ A cyclic prefix can null the interference due to time misalignments.
- ❑ Interference from the carrier frequency offsets cannot be totally compensated.

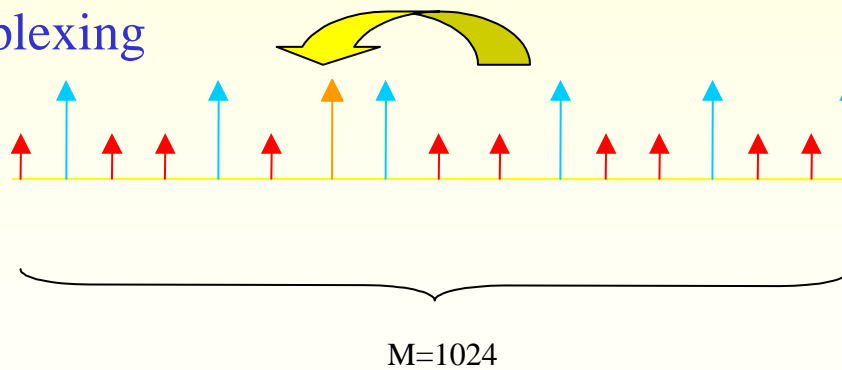
Ref: Tonello, Laurenti, Pupolin, Proc. *ICT 2000*, Proc. *VTC 2000 Fall*.

MU-OFDM: SIR on Adjacent Sub-Carrier

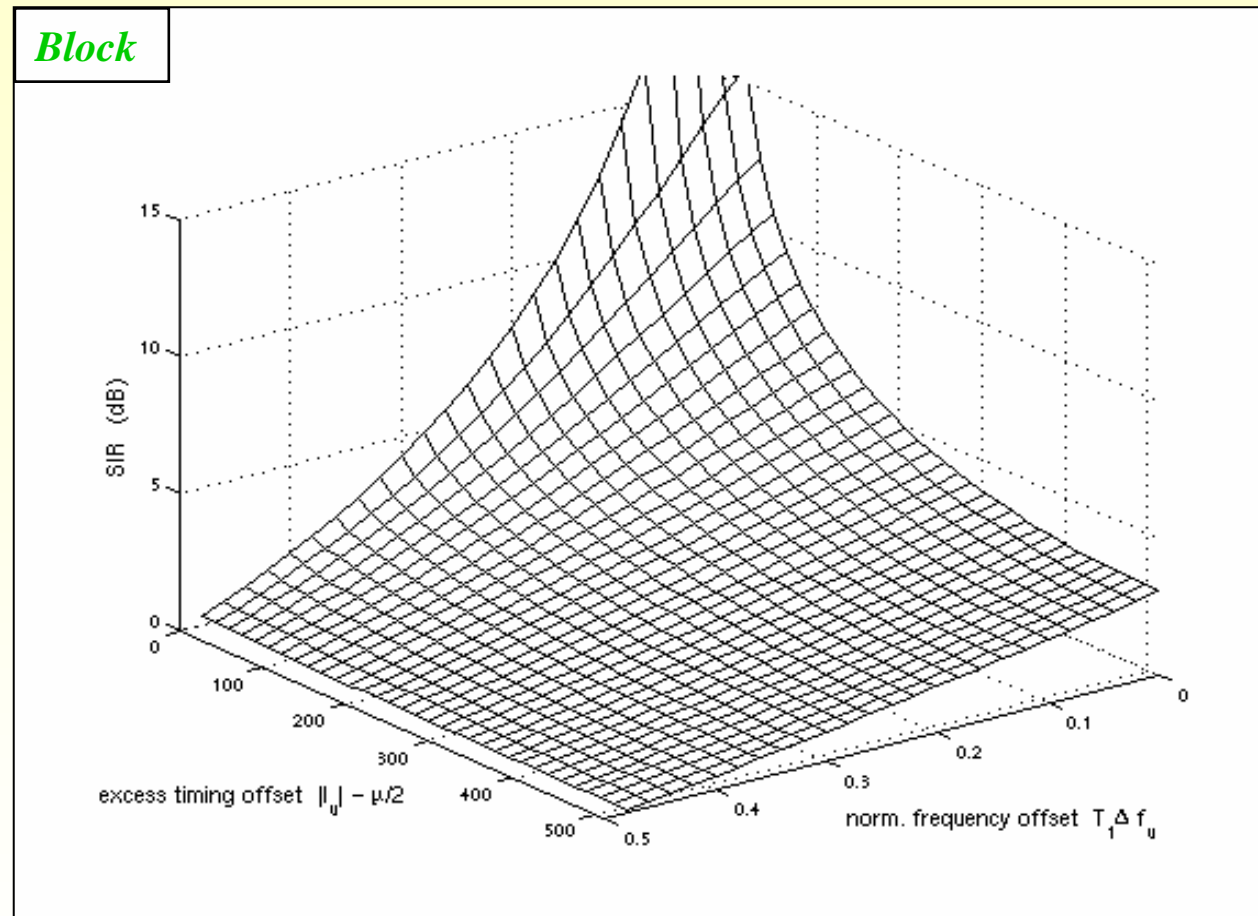
Block Multiplexing



Interleaved Multiplexing

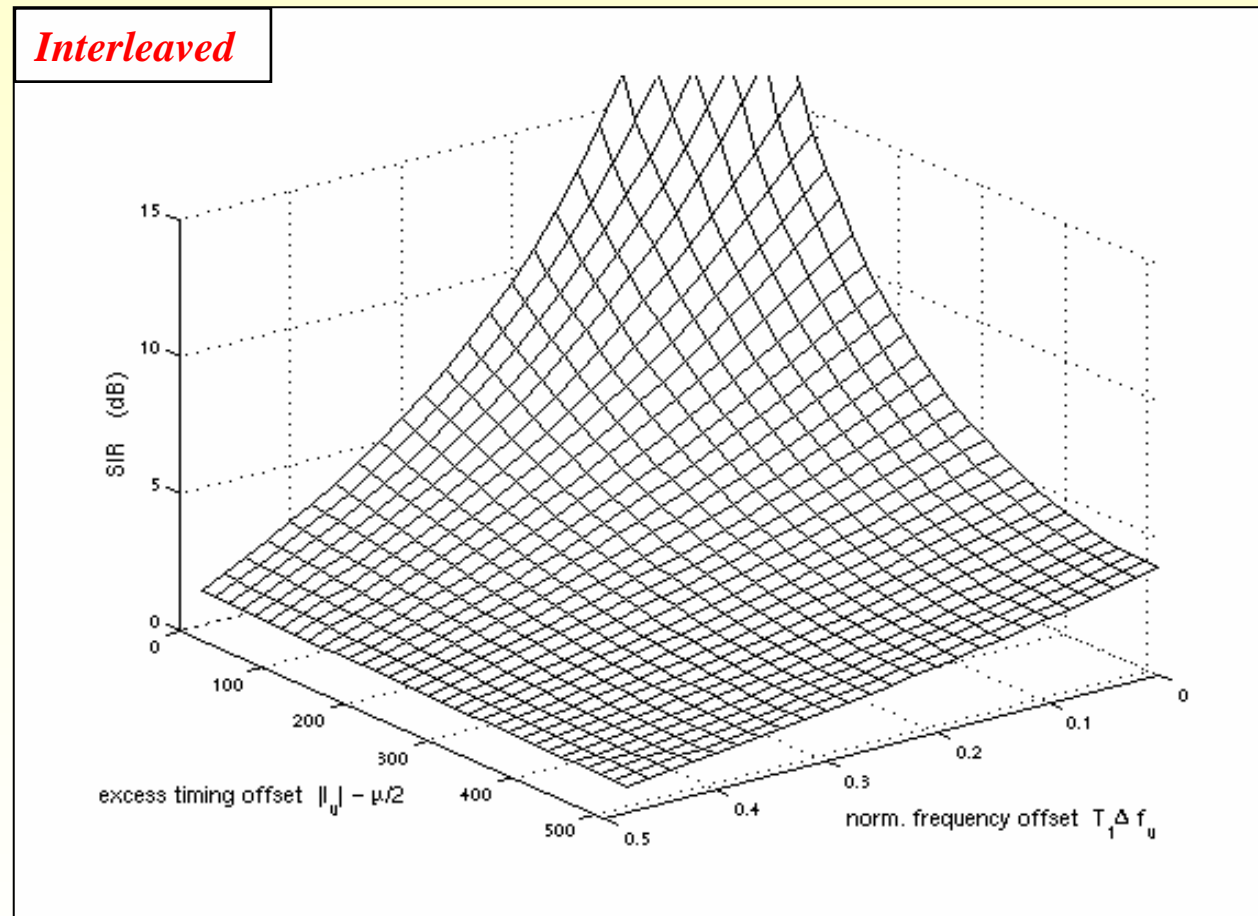


SIR with Block Allocation in MU-OFDM

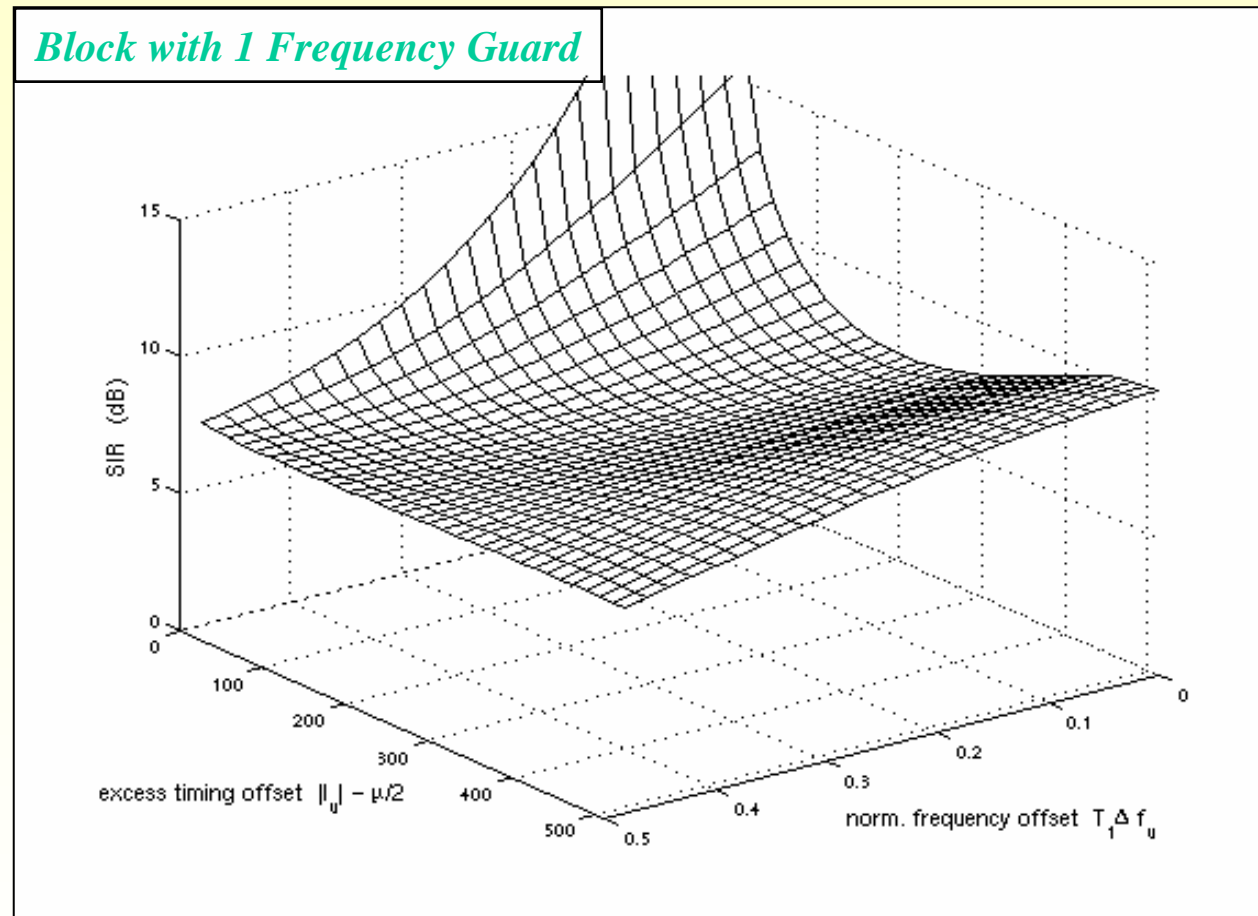


The SIR rapidly decreases as the frequency offset and the time offset increase.

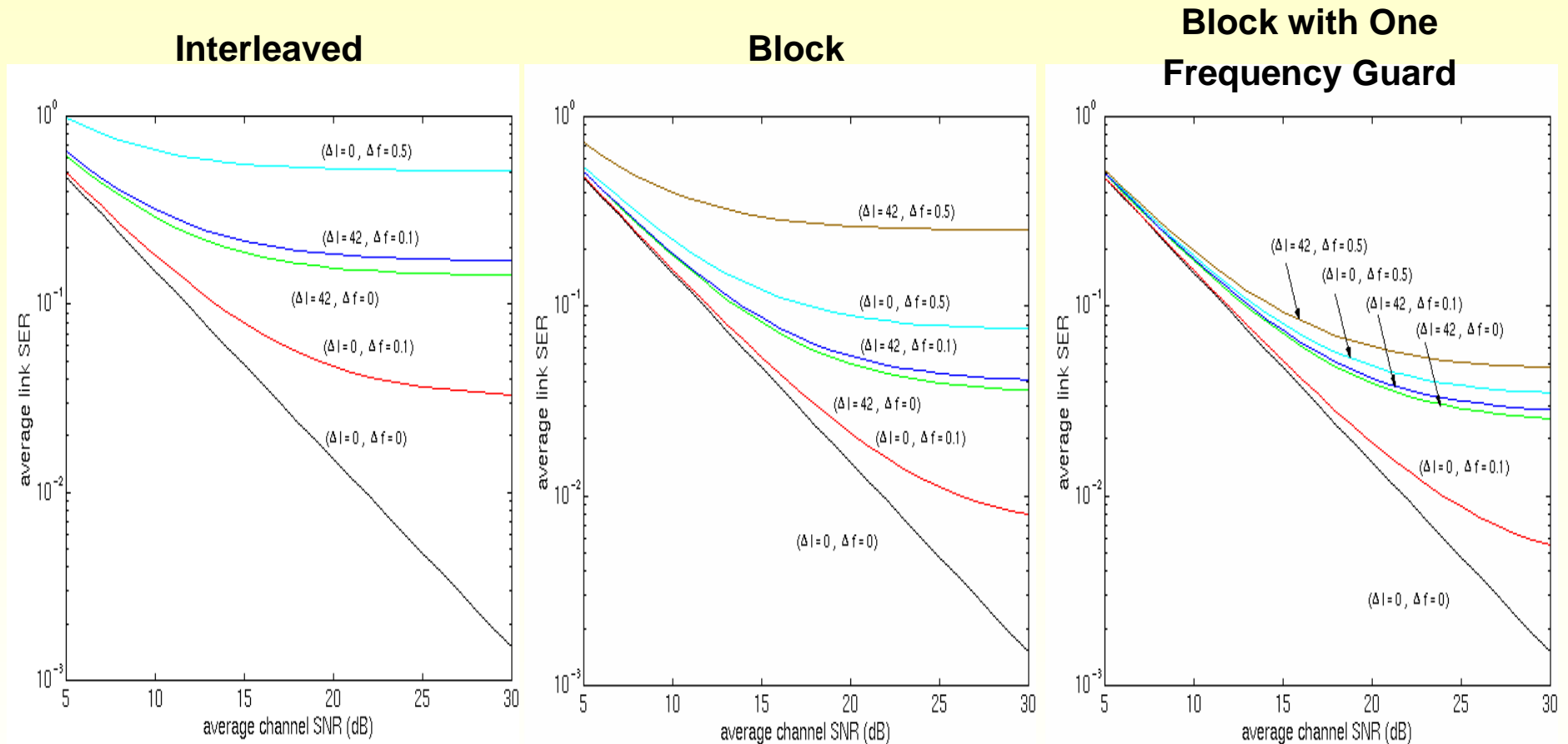
SIR with Interleaved Allocation in MU-OFDM



SIR with Block Allocation and 1 Guard in MU-OFDM



MU-OFDM: Average Symbol Error Rate



4-PSK, 16 users, 12 tones/user, $\mu=64$, $W=4.096$ MHz.

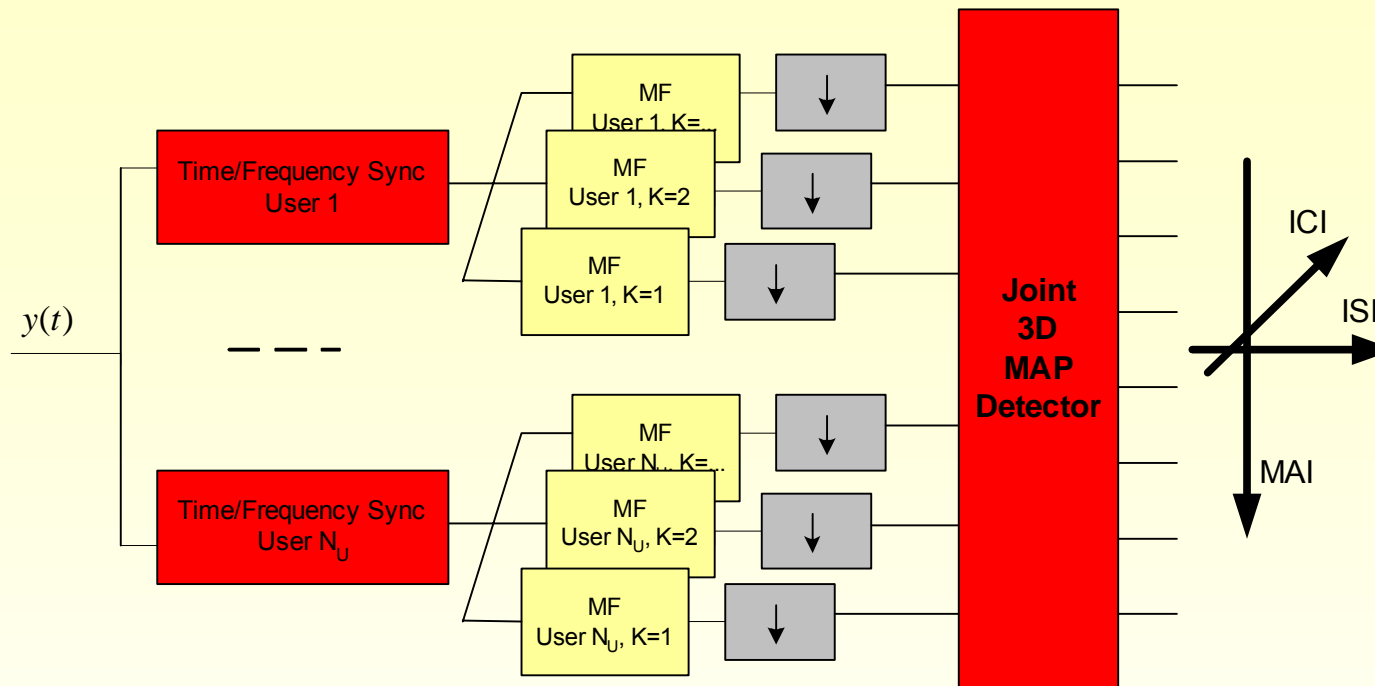
Excess time offsets and frequency offsets are independent and uniformly distributed between $[-\Delta l, +\Delta l]$ and $[-\Delta f, +\Delta f]$ respectively. The channel has an exponential power delay profile with d.s.=10 μ s.

Remarks

- ❑ MU-OFDMA is severely affected by users' time/frequency offsets in the uplink.
- ❑ Time/Frequency guards reduce the spectral efficiency.
- ❑ Acquisition of timing and carrier frequency is difficult.
- ❑ Due to the spectral containment characteristics, FMT is a better option than OFDM for the uplink.

Optimal Multiuser Multitone Detection

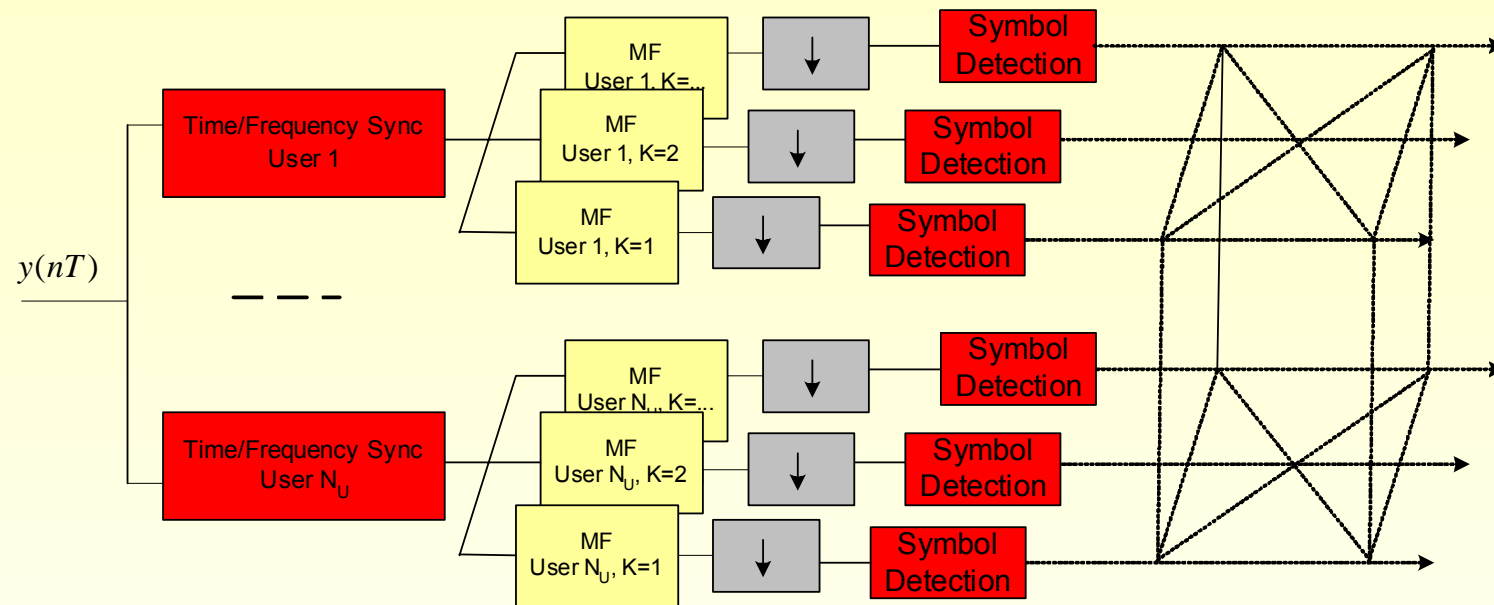
Optimal MLSE/MAP Detection



- Optimal MAP Detection has a complexity that grows exponentially with the number of users, tones, and sub-channel memory.
- With frequency concentrated pulses and orthogonal tone assignment it simplifies to a bank of single channel MAP detectors.

Ref: Tonello, *VTC 2002 Fall, Bell Labs Tech. Jour. 2003*

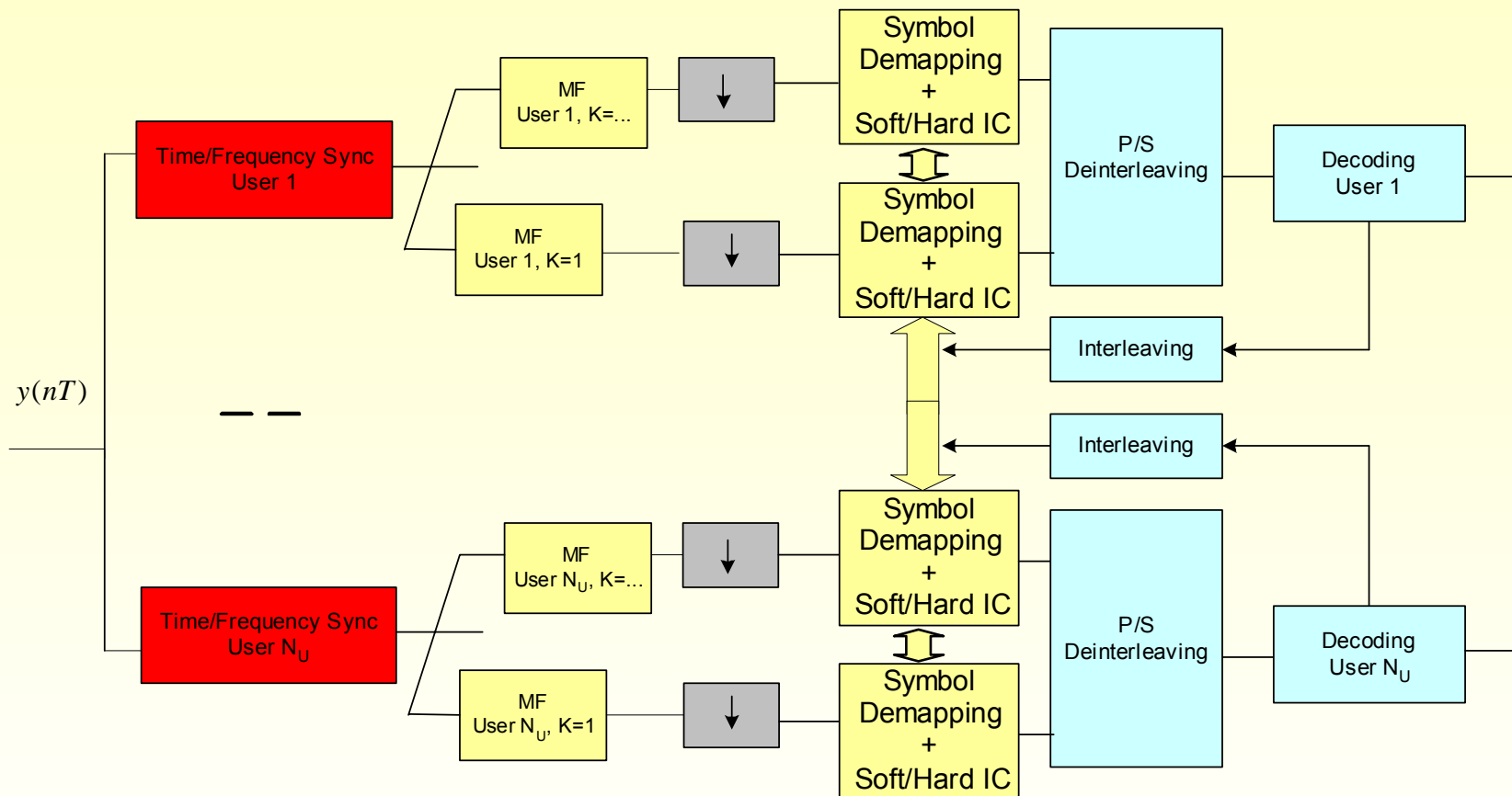
Turbo Per-Symbol Detection



Simple approach

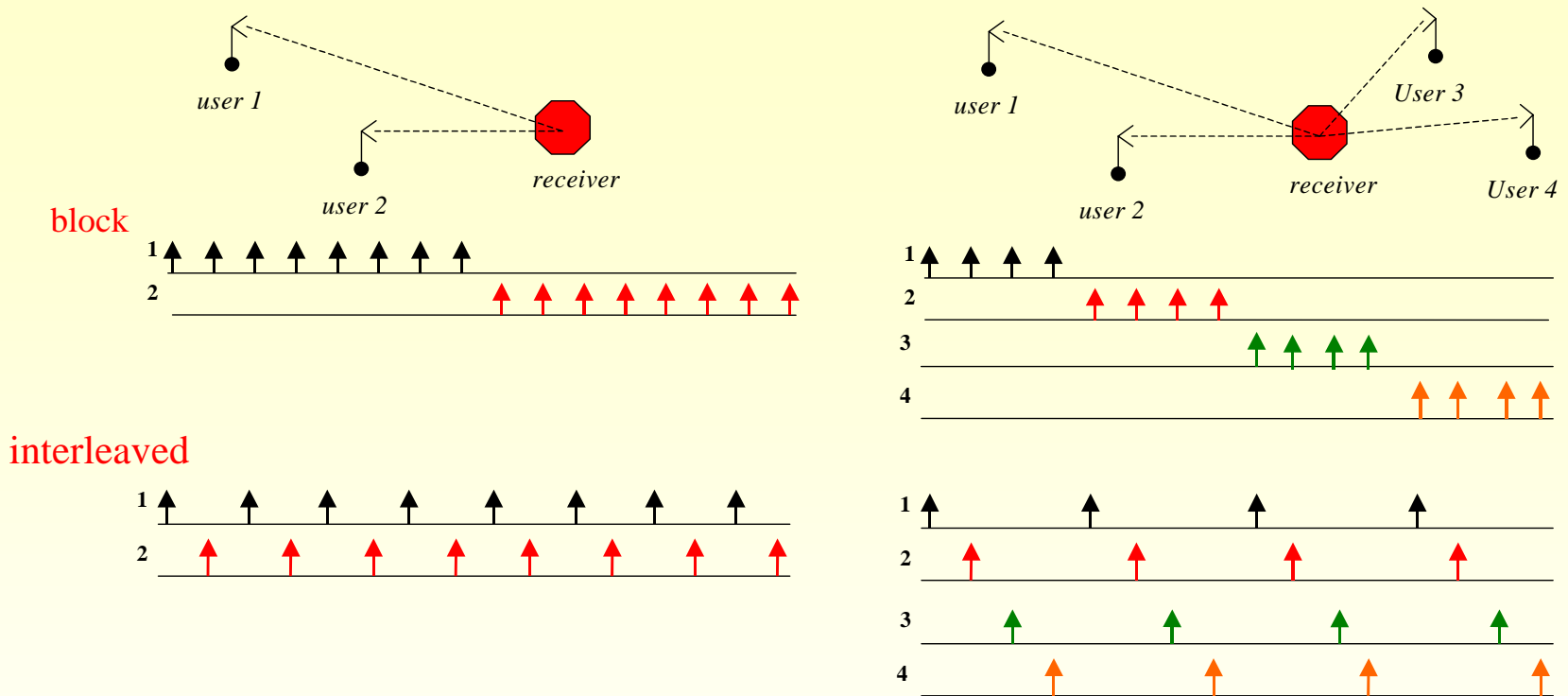
- Symbol-by-symbol detection with iterative interference cancellation

Turbo Per-Symbol Decoding



- Feedback from the decoders if we assume to use bit-interleaved codes

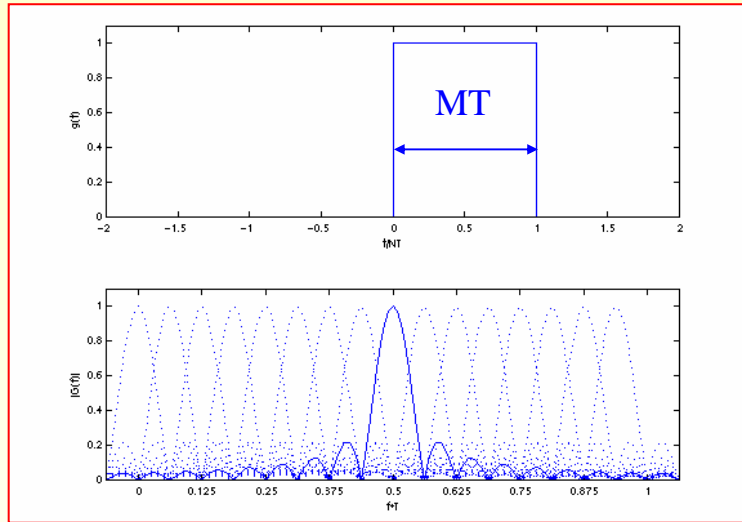
Example of Multiuser Multitone System



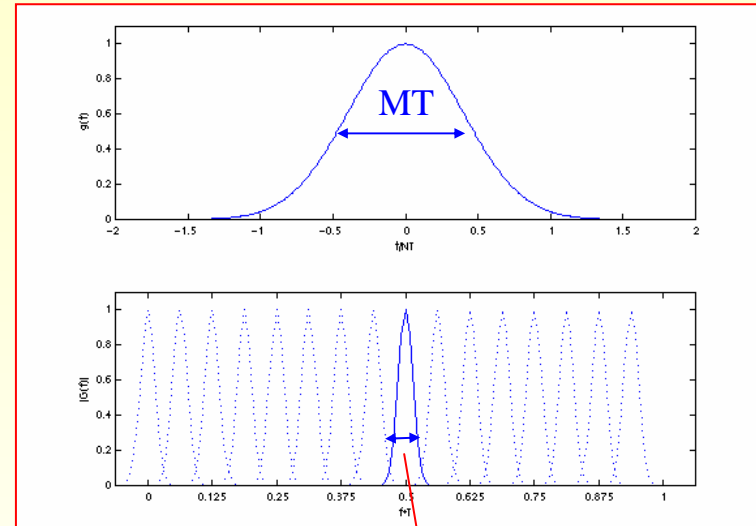
- 16 Total carriers
- QPSK – $\frac{1}{2}$ Convolutional Code
- Δt uniformly distributed in $[-\Delta t_m/2, \Delta t_m/2]$
- Δf uniformly distributed in $[-\Delta f_m/2, \Delta f_m/2]$

Sub-Channel Pulses

DMT: Rectangular Pulse



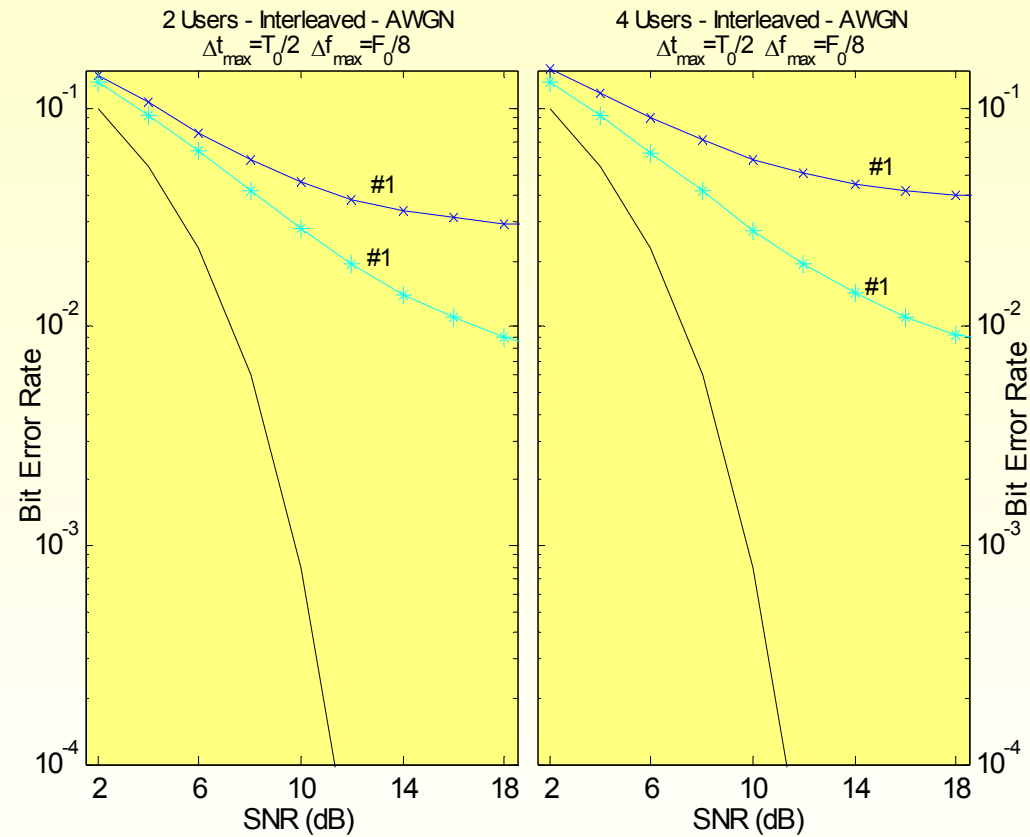
FMT: Gaussian Pulse



$1/(MT)$

- Minimal sub-carrier spacing (no cyclic prefix)

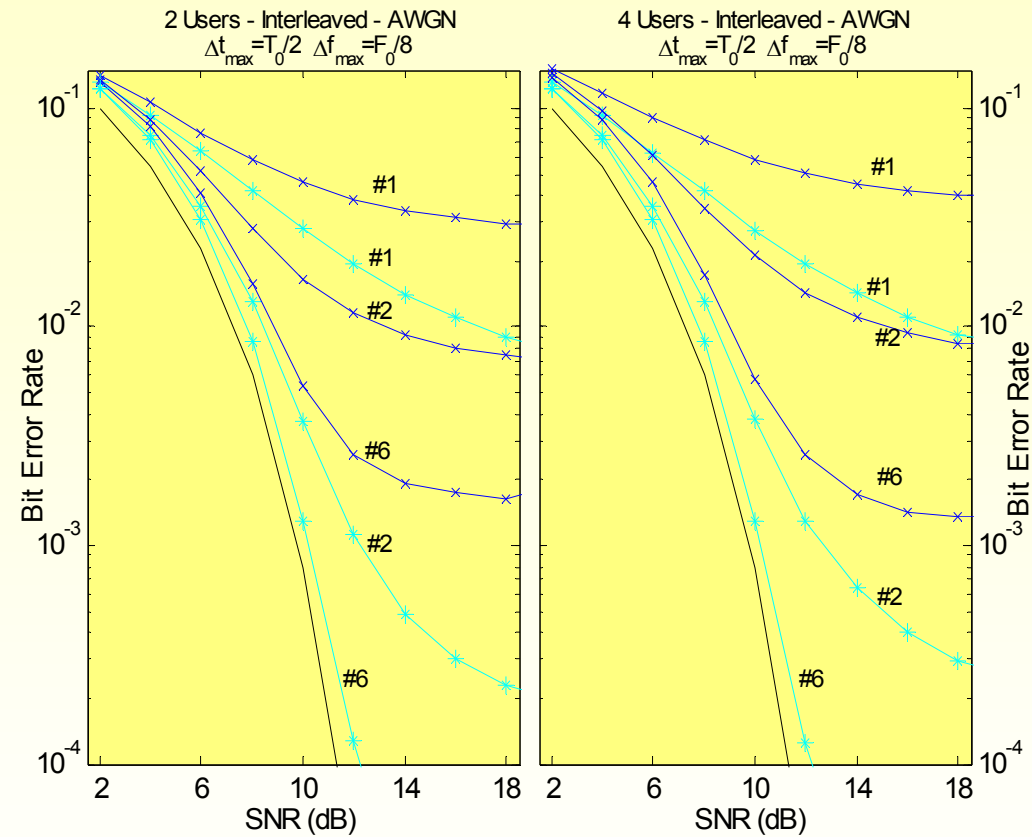
BER Unencoded - Interleaved



Uncoded

- RECT pulses
- GAUSS pulses

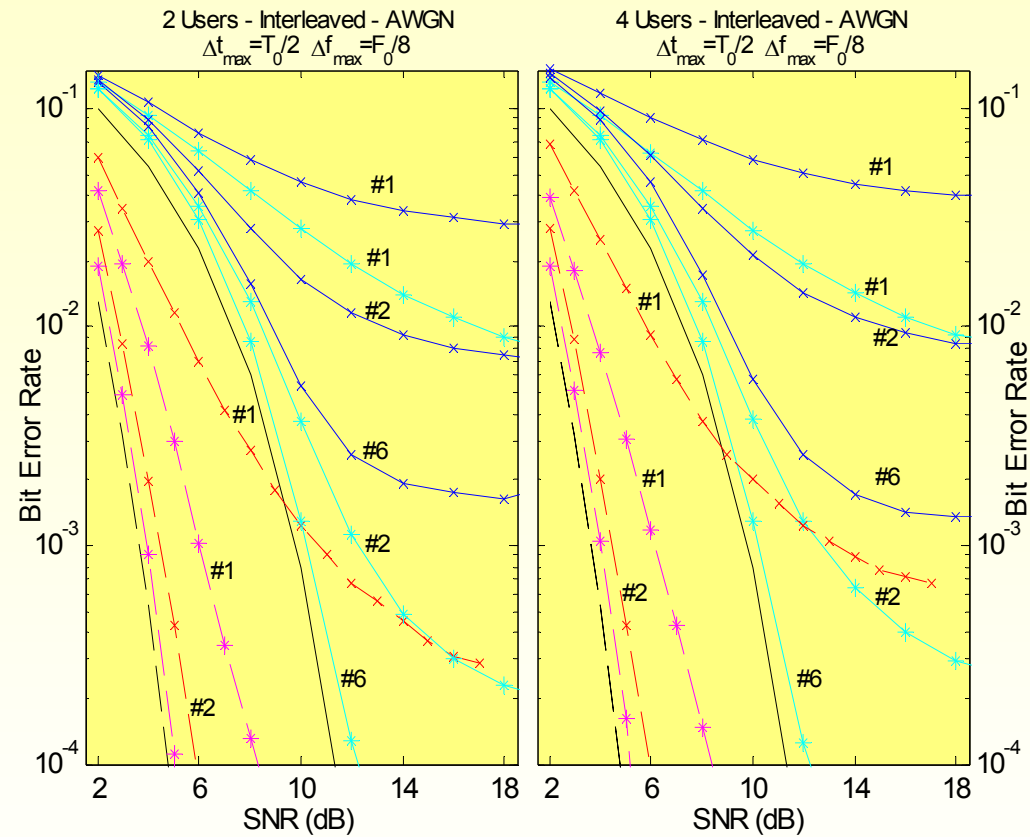
BER Unencoded - Interleaved



Uncoded

- RECT pulses
- GAUSS pulses

BER Coded - Interleaved



Remarks

- ❑ FMT is a better solution than DMT for uplink asynchronous communications:
 - ✓ Single user detection with sub-channel equalization works fine if we use appropriate frequency confined sub-channel pulses!
- ❑ Optimal performance can be achieved with multitone multiuser detection.
- ❑ Simple iterative per-symbol detection/decoding shows fast convergence to the matched filter bound.

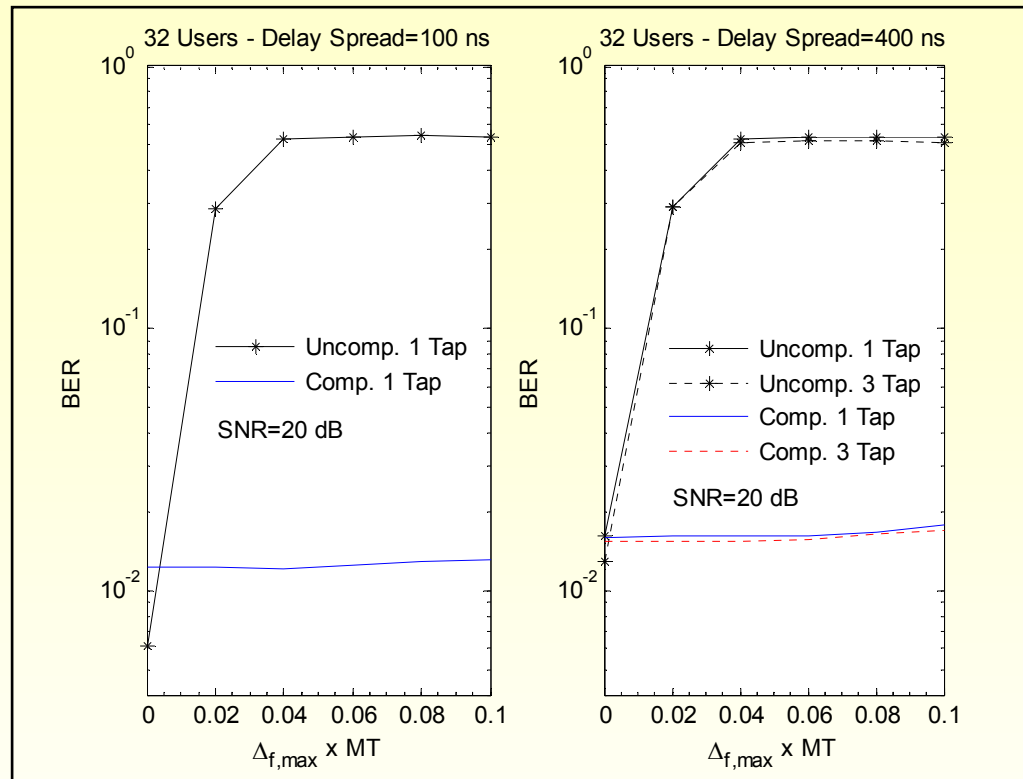
Synchronization in MU-FMT

MU-FMT Synchronization Algorithm

- We assume a single user receiver with training sequences and a 1-3 taps RLS equalizer per-sub-channel.

Ref: Tonello, Pecile, Proc. *VTC 2005 Spring*.

Multiuser FMT - Synchronization Performance



- 4-PSK Modulation, $M=32$ Sub-channels, Root-Cosine Pulses, Bandwidth 10 MHz , Frequency guards of 12.5 kHz. Training of length 15 symbols.
- Delay spread 100 ns and 400 ns.
- Users are time-asynchronous with a random phase, the frequency offsets are uniformly distributed in $[-\mathbb{E} f_{max}, \mathbb{E} f_{max}]$. They have a single sub-channel each.

Remarks

- ❑ The practical scheme with training sequences performs well.
- ❑ The MU-FMT architecture is very robust to the multiple access interference.
- ❑ Single user detection is simply required.

Final Conclusion

FMT is a promising technology that has great potentiality for the broadband uplink scenario !

Some References

Multicarrier Systems Performance Analysis

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